

$$H(z) = H_a(s) \Big|_{s = \frac{z-1}{T_s}} = \frac{1-z^{-1}}{1+z^{-1}}$$

EX: Transform

$$H_a(s) = \frac{s+1}{s^2+5s+6}$$

Into a digital filter using the bilinear transformation method. Choose $T_s = 1$

SOL:

$$H(z) = \frac{2 \left(\frac{1-z^{-1}}{1+z^{-1}} \right) + 1}{4 \left(\frac{1-z^{-1}}{1+z^{-1}} \right)^2 + 5 \cdot 2 \left(\frac{1-z^{-1}}{1+z^{-1}} \right) + 6}$$

$$= \frac{3 + 2z^{-1} - z^{-2}}{20 + 4z^{-1} + 20z^{-2}}$$

$$\therefore H(z) = \frac{\overset{b_0}{0.15} + \overset{b_1}{0.1} z^{-1} - \overset{b_2}{0.05} z^{-2}}{1 + \underset{a_1}{0.2} z^{-1} + \underset{a_2}{1} z^{-2}}$$

$$y[n] + 0.2 y[n-1] + y[n-2]$$

$$= 0.15 x[n] + 0.1 x[n-1] - 0.05 x[n-2]$$

EX: Design a digital filter with the following specs. using BTM.

$$0.89125 \leq |H(e^{j\omega})| \leq 1; 0 \leq \omega \leq 0.2\pi$$

$$|H(e^{j\omega})| \leq 0.178; 0.3\pi \leq \omega \leq \pi$$

SOL:

$$\Omega = \frac{Z}{T_s} \tan\left(\frac{\omega}{2}\right)$$

$$T_s = 1 ; \text{ for convenience.}$$

\therefore the corresponding analog filter specs will be

$$0.89 \leq |H_a(j\omega)| \leq 1$$

$$0 \leq \Omega \leq Z \tan\left(\frac{0.2\pi}{2}\right)$$

$$|H_a(j\Omega)| \leq 0.178$$

$$Z \tan\left(\frac{0.15\pi}{2}\right) \leq \Omega \leq \infty$$

Because of monotonic response,

$$H_a(jZ \tan 0.1\pi) = 0.89$$

$$H_a(jZ \tan 0.15\pi) = 0.178$$

We know

$$|H_a(j\Omega)|^2 = \frac{1}{1 + \left(\frac{\Omega}{\Omega_c}\right)^{2N}}$$

$$\therefore 1 + \left(\frac{2 \tan(0.1\pi)}{\Omega_c} \right)^{2N} = \frac{1}{0.89^2} \quad (1)$$

$$1 + \left(\frac{2 \tan(0.15\pi)}{\Omega_c} \right)^{2N} = \frac{1}{0.178^2} \quad (2)$$

Solving (1) & (2)

$$N = \frac{\log \left[\frac{\frac{1}{0.178^2} - 1}{\frac{1}{0.89^2} - 1} \right]}{2 \log \left[\frac{\tan 0.15\pi}{\tan 0.1\pi} \right]}$$

$$= 5.305 \approx 6 \quad \leftarrow \begin{array}{l} \text{must} \\ \text{be next} \\ \text{highest} \\ \text{integer.} \end{array}$$

Putting $N=6$ into equ (2) we get $\Omega_c = 0.766$. For this value of Ω_c the pass band specs are met exactly. This is reasonable with the BTM, as we don't have to be concerned with aliasing.

Similar to example 7.2 to get $H_a(s)$

$$H_a(s) = \frac{0.2024}{(s^2 + 0.3995s + 0.5871)(s^2 + 1.0836s + 0.5871)(s^2 + 1.4802s + 0.5871)}$$

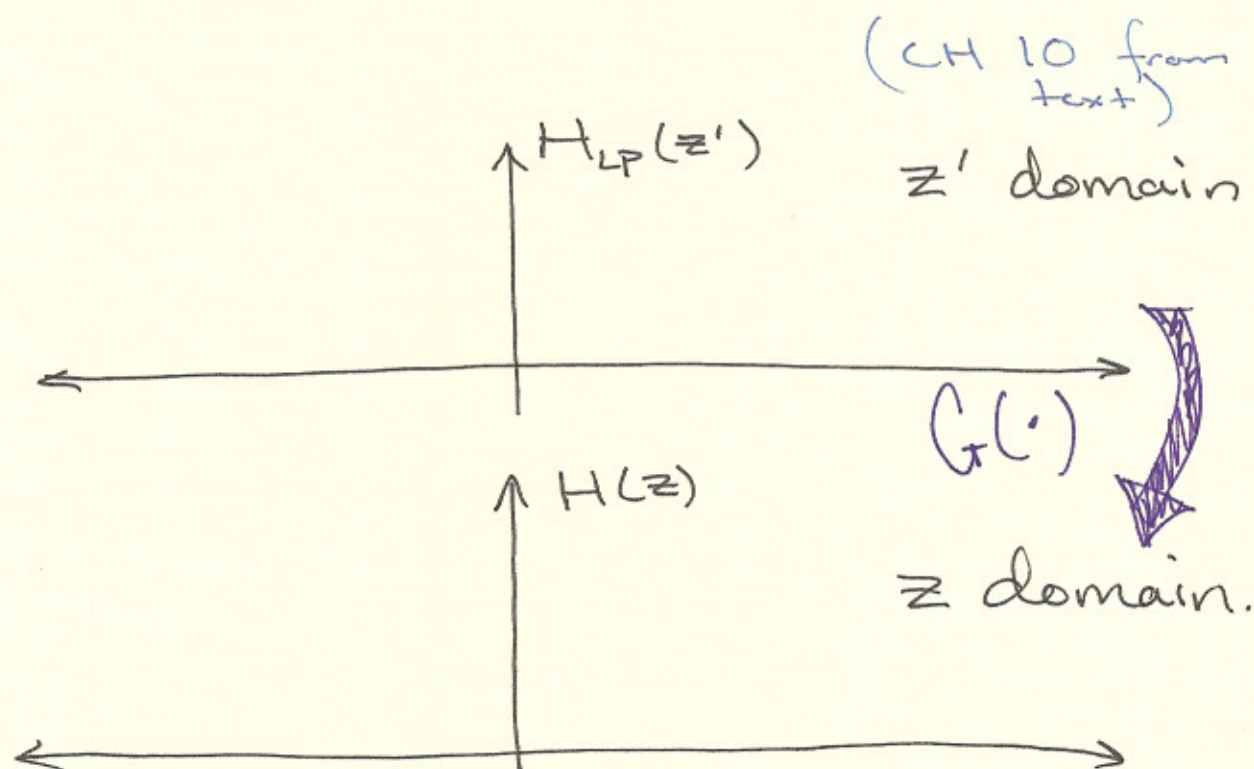
$$H(z) = H_a(s) \Big|_{s = \frac{z-1}{T_s} \frac{1+z^{-1}}{1-z^{-1}}}$$

* use Butterworth polynomial table to get $H_a(s)$.

To create a different type filter, we create a LP prototype, then we go to the transformation table.

4.3 Frequency Band Transformation

Design of H.P. B.P. and B.S. filters from a prototype lowpass filter



$$z' = e^{j\omega'}$$

$$z = e^{j\omega}$$

let $H_{LP}(z')$ is TF of the prototype filter
 $H(z)$ is " " " desired "

Assume that $H_{LP}(z')$ is a causal and stable filter. We want $H(z)$ to be stable and causal. This requires the following conditions

- ① The unit circle in z' plane must map onto the unit circle of the z plane.
- ② The ^{inside} unit circle in z' plane must map onto the inside of the circle in z -plane.

According to ①

$$|z^{-1}| = |G(z^{-1})| = |G(e^{j\omega})| = 1$$

$$\begin{aligned} e^{-j\omega'} &= G(e^{-j\omega}) = |G(e^{j\omega})| \angle G(e^{j\omega}) \\ &= |G(e^{-j\omega})| e^{j \angle G(e^{j\omega})} \end{aligned}$$

$$\therefore -\omega' = \angle G(e^{j\omega})$$

Based on these relations, the transformation. The general form of the function $G(\cdot)$ that satisfy all of the above requirements is a rational function of all pass types given by

$$z'^{-1} = G(z^{-1}) = \pm \prod_{k=1}^n \left(\frac{z^{-1} - \alpha_k}{1 - \alpha_k z^{-1}} \right)$$

Where, $|\alpha_k| < 1$ for stability req. ② and α_k will vary depending on the

type of desired filter

Handout \rightarrow Table 8.2

Text \rightarrow Table 10.2

★ We should know LP to LP and LP to HP. ★

EX: Consider the system function of a prototype LP filter is given as $H_{LP}(z')$

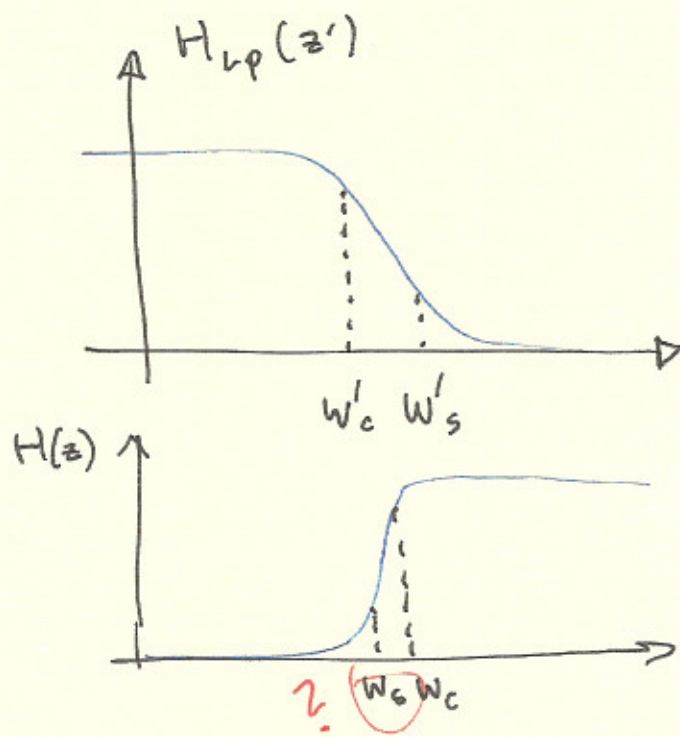
$$H_{LP}(z') = \frac{0.001836 (1+z'^{-1})^4}{(1 - 1.5z'^{-1} + 0.848z'^{-2})(1 - 0.155z'^{-1} + 0.6493z'^{-2})}$$

and $\omega'_c = 0.2\pi$

$\omega'_s = 0.3\pi$

Design a HP filter with the same tolerance but cut off freq at $\omega_c = 0.6\pi$

Sol:



Using table 8.2....

$$K = - \frac{\cos\left(\frac{\omega'_c + \omega_c}{2}\right)}{\cos\left(\frac{\omega'_c - \omega_c}{2}\right)}$$

$$= -0.382$$

$$H(z) = H_{LP}(z') \bigg|_{z'^{-1} = \frac{z^{-1} - 0.382}{1 - 0.382z^{-1}}}$$

$$H(z) =$$